



A 3ps Cryogenic Time-to-Digital Converter for Time-Correlated Single Photon Counting

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Better TDC Precision Looks Like...

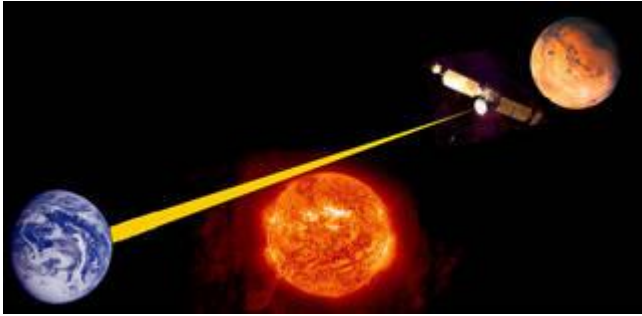


Image Credit NASA [1]

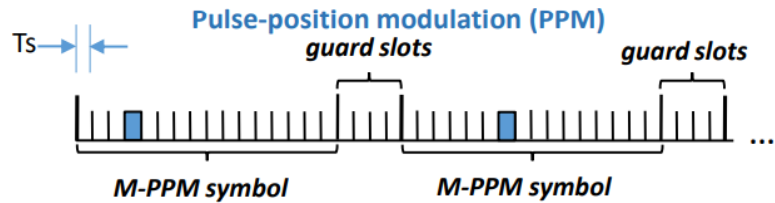
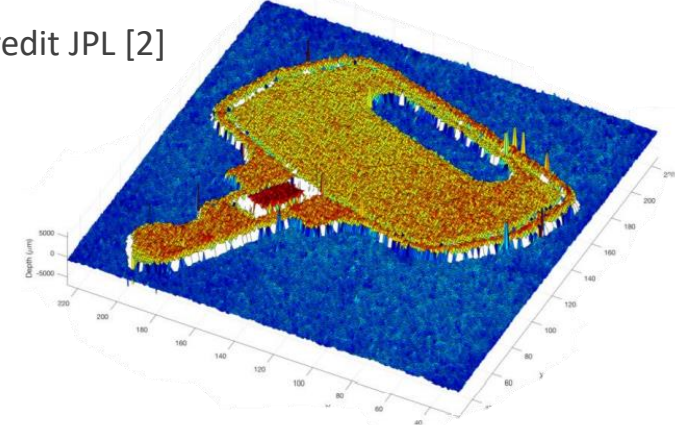


Image Credit JPL [2]

... higher bit-rates in DSOC.



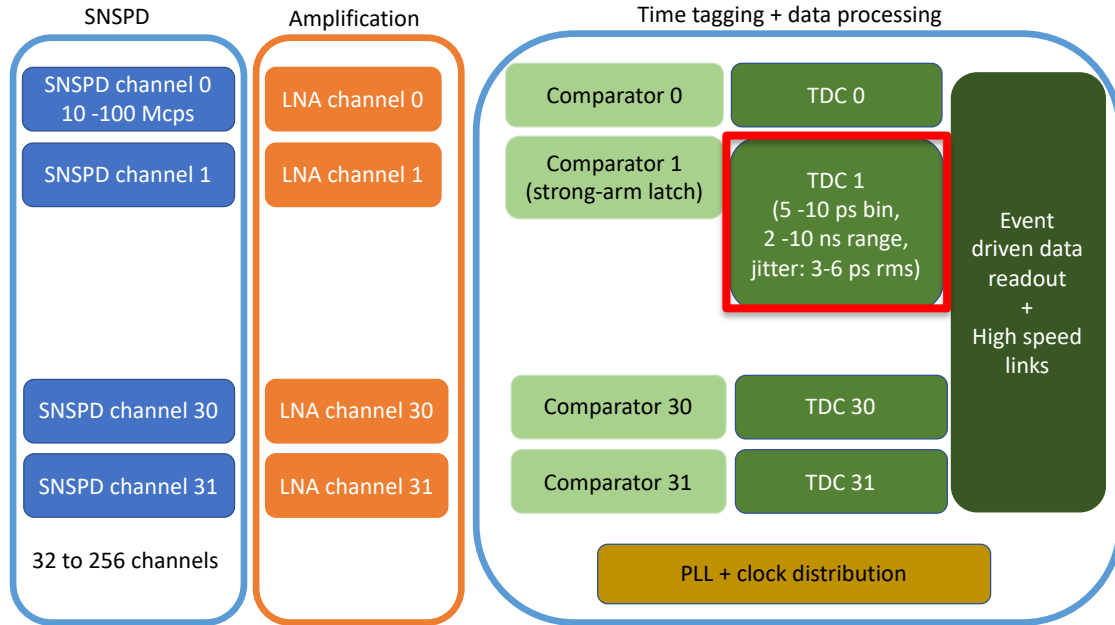
Image Credit JPL [2]



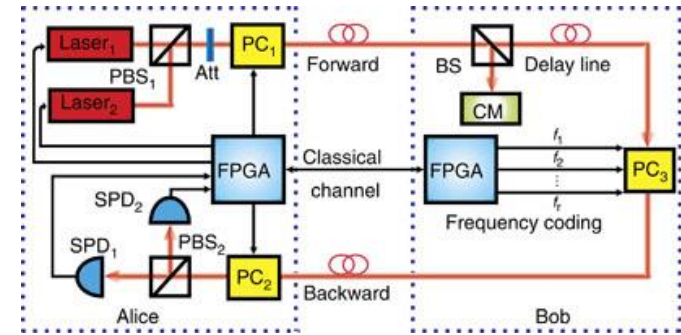
... ultra-high resolution range-finding.
(1 ps \approx 0.3 mm)

Why (This) TDC?

The Challenge: High-bandwidth time-correlated single photon counting in a cryogenic environment.



DOE NASA RFI 86 FR6315 [4]

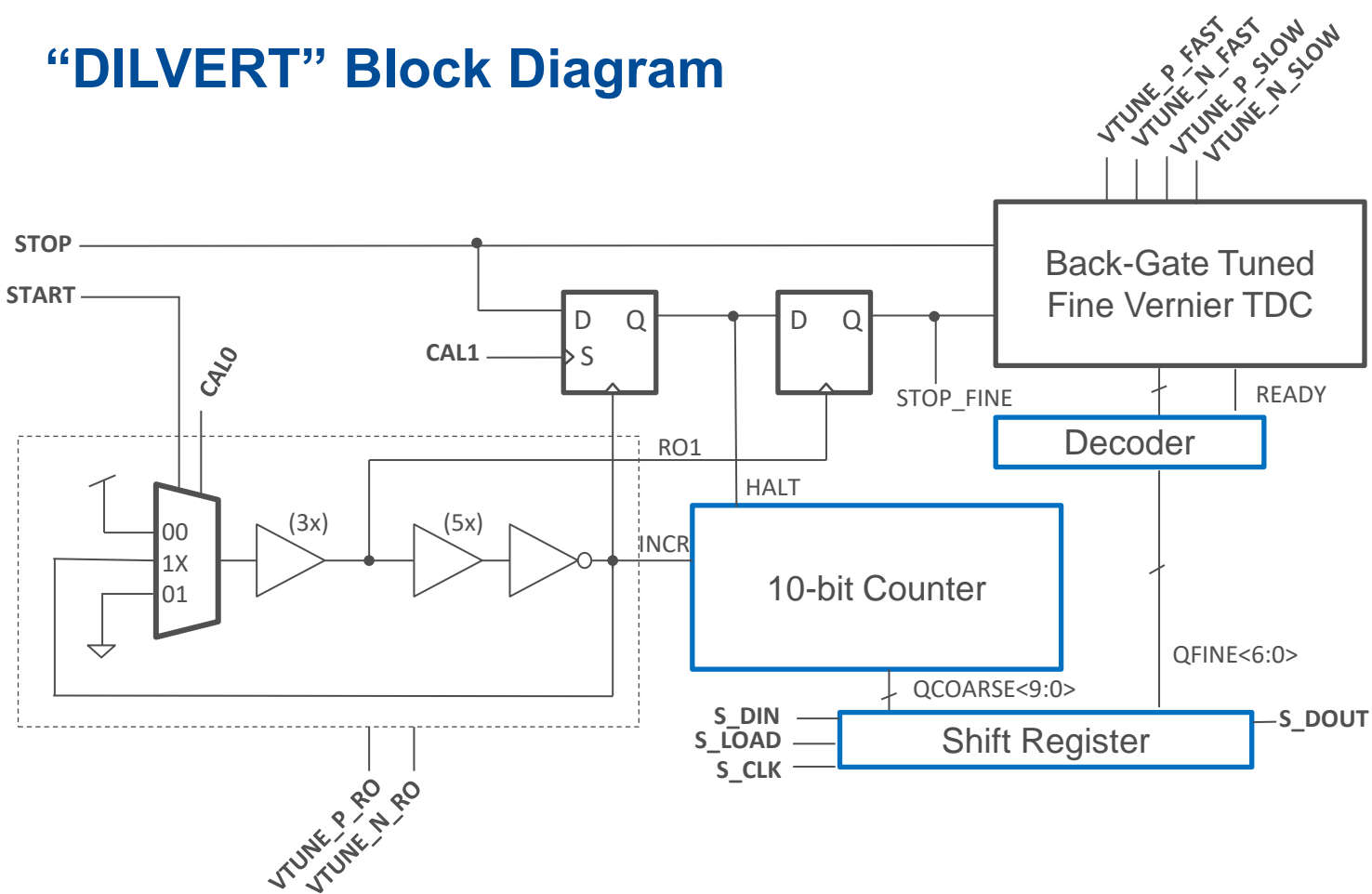


A Quantum Secure Direct Communications System (Hu [10])

1 – 10 Gcps

Proposed Application:
SNSPD-based space-to-ground quantum communications terminal for the ISS

“DILVERT” Block Diagram



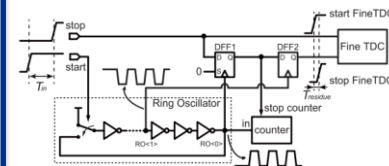
Total Pins: 22

- 6 analog inputs
- 7 digital inputs
- 3 digital outputs
- 6 power pins

Not Shown:

* RO1, HALT, INCR, and STOP_FINE are mux'd to **two** digital outputs with the select done by 4 shift register bits.

* RESET pin



Enomoto et al. [2]

Fine Vernier TDC

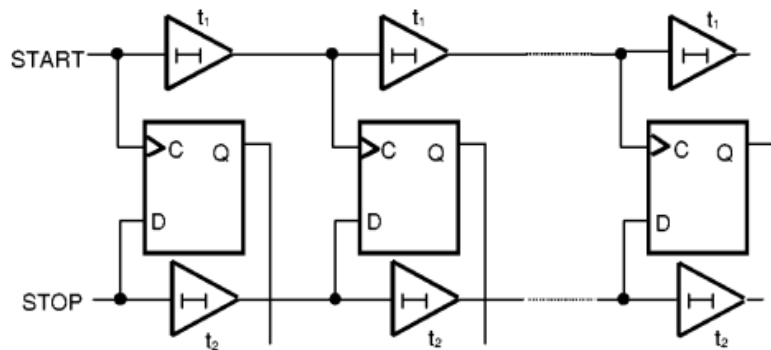
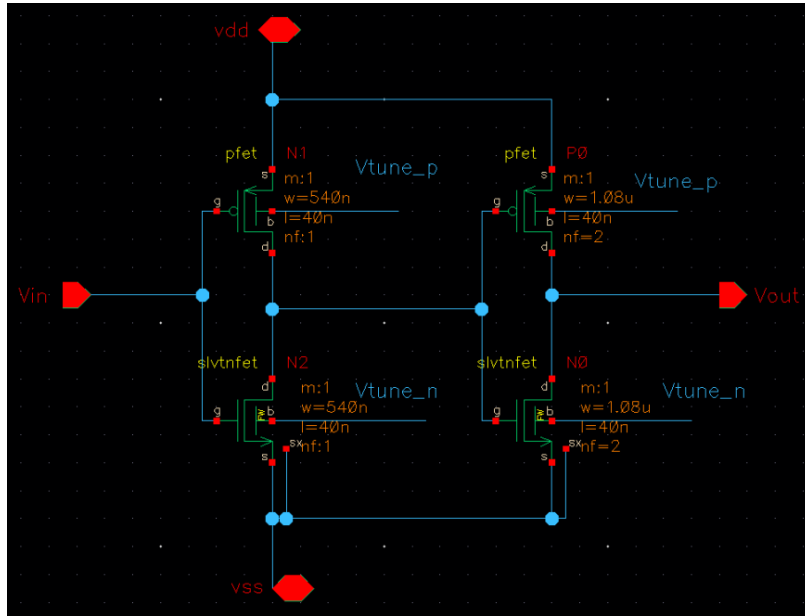


Illustration from Dudek [1]
(in our TDC, latches are used
instead of DFFs)

- t_1 and t_2 are tuned to be very close to each other.
- $1 \text{ LSB} = t_{R(\text{fine})} = (t_2 - t_1)$
- STOP is always faster than START.
- If START arrives first at stage n , $Q[n]=1$
- Result is thermometer-coded: “How many LSBs does it take for STOP to catch up to START?”

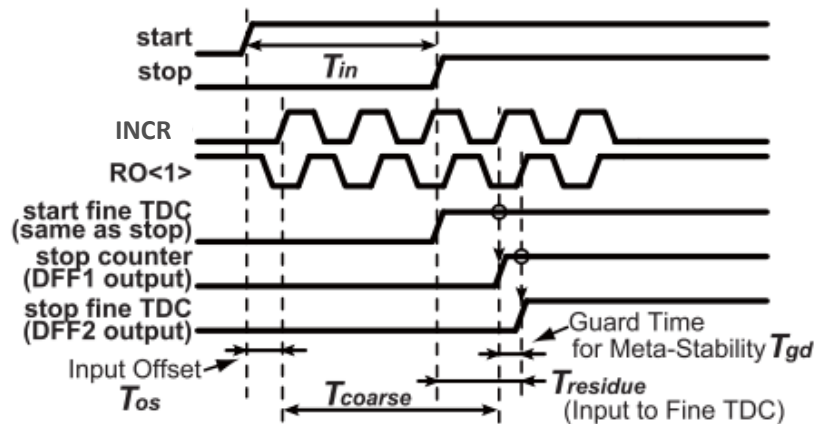
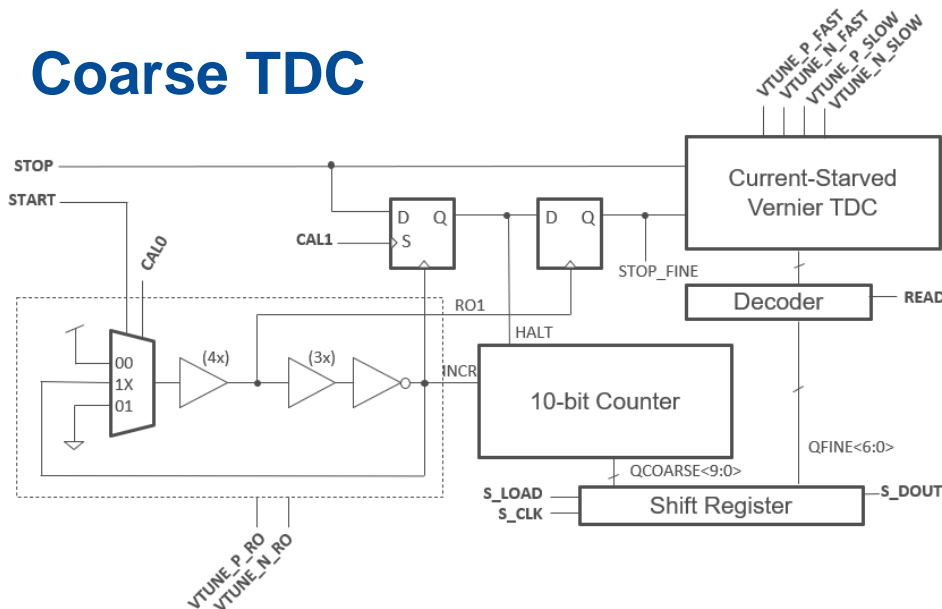
Delay Tuning with Back-Gate Bias



Fine TDC Delay Stage

- Fine TDC delay cells and RO delay cells are same architecture (different sizing)
- Speed is tuned by FDSOI back-gate biasing
 - Low complexity
 - Small footprint
 - No noise contribution from bias circuits
- 6 separate tuning voltages:
 - Fine TDC fast chain, slow chain, and coarse TDC ring oscillator
 - Tune NMOS and PMOS separately.
 - Both 0~2V w.r.t VSS

Coarse TDC

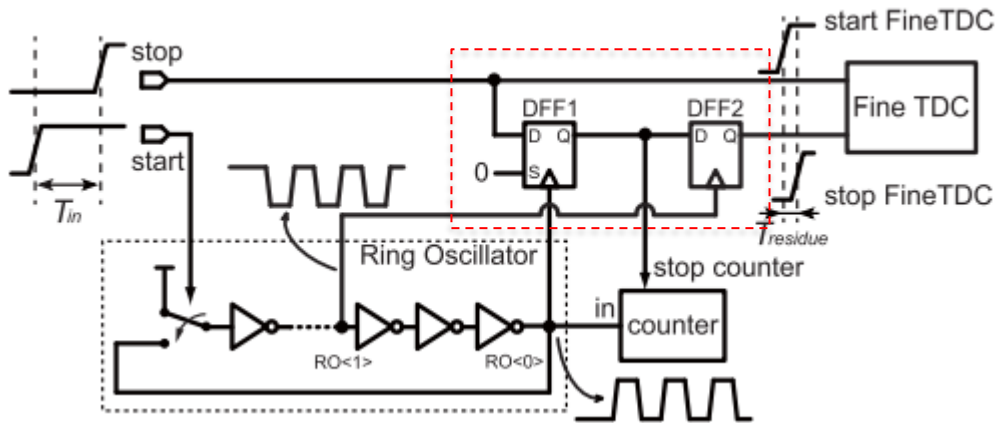


Enomoto et al. [2]

START starts the ring oscillator. The counter counts up until STOP, and the residue is encoded by the fine TDC:

$$\begin{aligned} T_{in} &= T_{OS} + T_{coarse} + T_{gd} - T_{residue} \\ &= (T_{OS} + T_{gd}) + N_{coarse}(t_{R(coarse)}) - N_{fine}(t_{R(fine)}) \end{aligned}$$

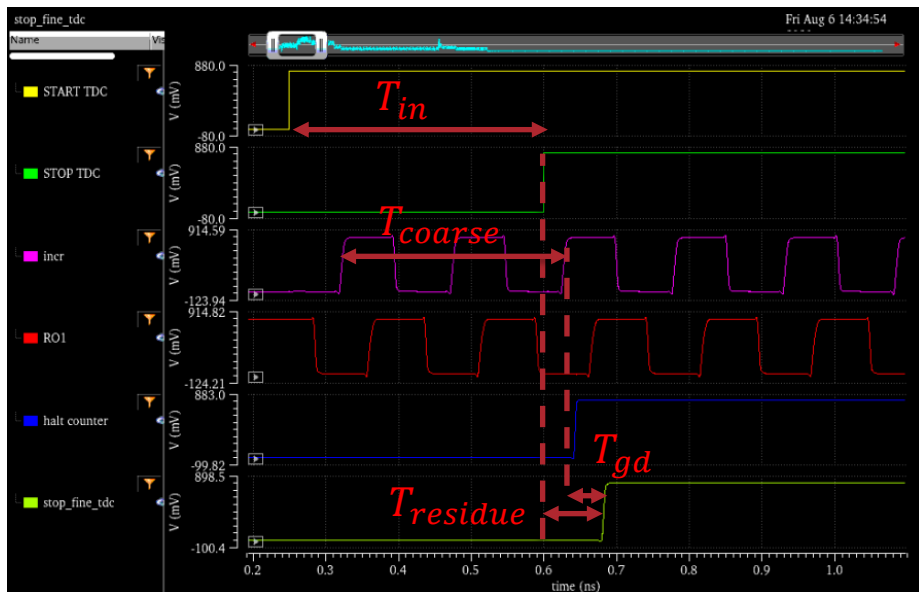
Coarse-to-Fine Coupling with DFFs



Enomoto, et al. [2]

- Traditional TDCs couple with a multiplexer
→ unwanted $T_{D \rightarrow Q}$, possibly different for different inputs. ☹
- In our design, DFF1 is triggered on the *first coarse TDC increment after STOP arrives*.
(With constant offset $T_{clk \rightarrow Q}$)
- The remainder ($t_{incr} - t_{STOP}$) is digitized by the fine TDC.
- Why have DFF2? STOP and RO<0> are asynchronous, so we need to suppress any metastability in DFF1's output before the fine TDC.
- DFF2 is triggered slightly after DFF1 ($T_{incr} + T_{gd}$). We assume DFF1's output has settled within T_{gd} .

Example Measurement



From Three Point Calibration, we have:

$$t_{R(fine)} = -2.6375 \text{ ps}$$

$$t_{R(coarse)} = 152.75 \text{ ps}$$

$$T_{OS} + T_{gd} = -173.62 \text{ ps}$$

For this measurement, $Q_{fine} = 33$ and

$Q_{coarse} = 4$, so:

$$T_{in} = T_{OS} + T_{coarse} + T_{gd} - T_{residue} \\ \approx 350.343 \text{ ps}$$

(Actual T_{in} was 350 ps, within 1 LSB of the measured value.)

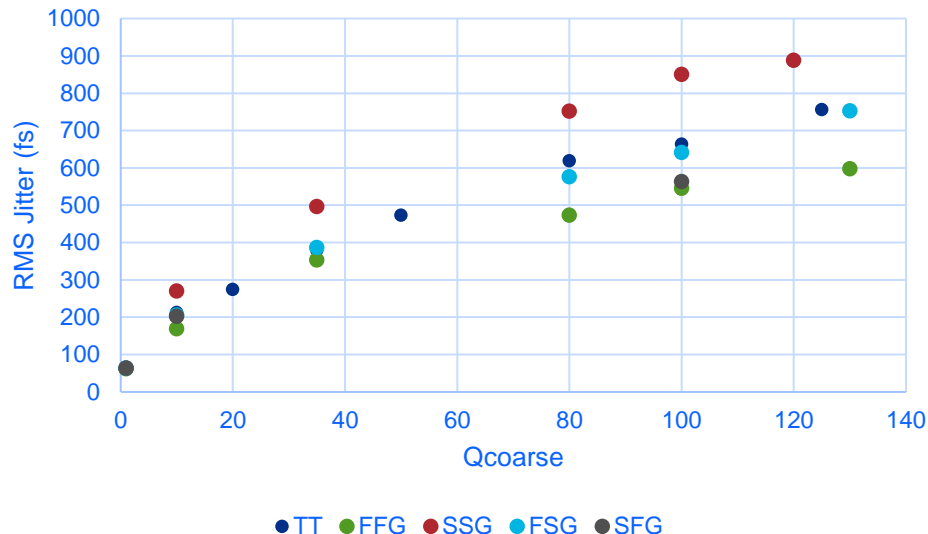
Simulation Results: TDC Jitter vs T_{in}

$$\begin{aligned}\sigma_{jitt} &= \sigma_{tp} \times \sqrt{N_{stage} \times 2 \times \sqrt{Q_{coarse}}} \\ &= \kappa_{jitt} \sqrt{Q_{coarse}}\end{aligned}$$

Assuming that maximum allowable jitter is approximately $\frac{1}{2}$ LSB:

Corner:	TT	FF	SS	FS	SF
κ_{jitt} [fs]	66.18	55.9	80.7	65.0	61.7
$\sigma_{jitt(max)}$ [fs]	1.3	1.0	1.8	1.4	1.2
$Q_{coarse(max)}$	385	320	497	463	378
Dynamic Range [ns]	<u>55.4</u>	40.2	83.7	68.5	53.1

Ring Oscillator Time Jitter vs. Qcoarse



State of the Art

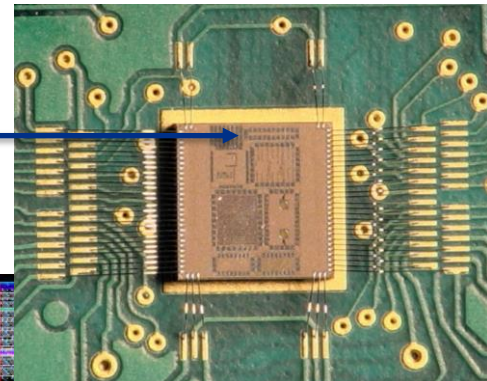
	Chen '17	Sesta '18	Enomoto '19	Palaniappan '19	Khaki '21	Lee '21	DILVERT (sim) '21
Architecture	$\Delta\Sigma$ TDC	Counter + Vernier Delay Line	Counter + Pulse-Shrink Delay Line	Cap Boost Vernier Delay Line	$\Delta\Sigma$ TDC w/ GSRO	SAR TDC	Counter + Vernier Delay Line
Process	65nm	0.35 μm	0.18 μm	180 nm	40nm FPGA	130nm	22nm SOI
Resolution	0.48 ps	7 ps	2.0 ps	1.74 ps	0.18 ps	6.6 ps	2.7 ps
Dynamic Range	-	80 ns	120 ns	0.112 ns	4.5 ns	1.7 ns *	55.4 ns
Power	3 mW	-	18 mW	0.217 mW	9.24 mW	0.504 mW	0.5 mW
Rate	-	20 MS/s *	3.3 MS/s	50 MS/s	1600 MS/s	10 MS/s	100 MS/s
FoM	-	-	3.33	0.30	2.71	0.51	41.04

$$FoM = \frac{[\text{Dynamic Range in ns}]}{[\text{Power in mW}][\text{Resolution in ps}]}$$

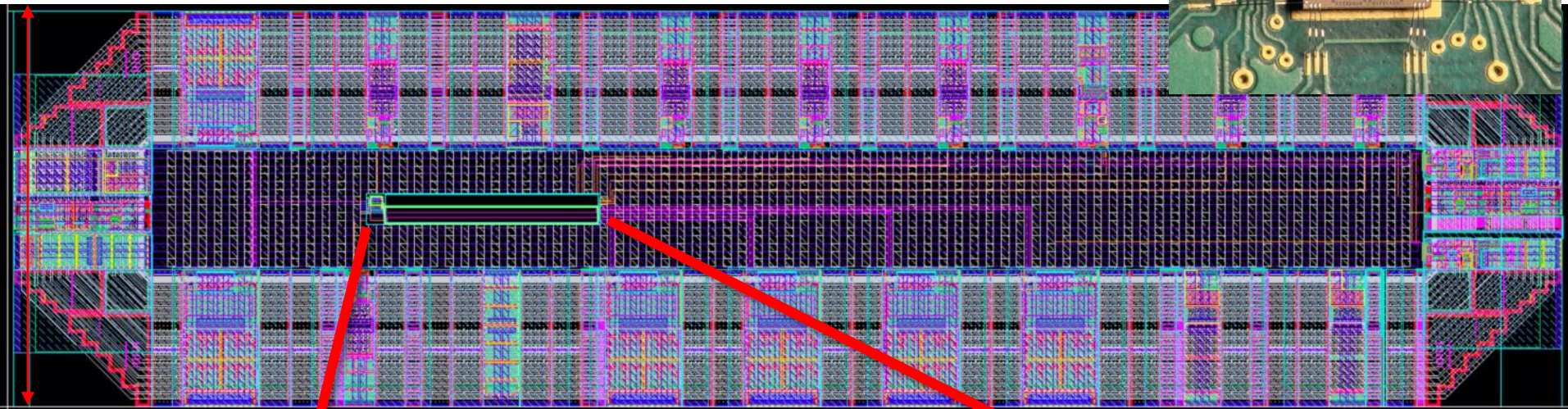
* = calculated from available data

Test Chip 11/2021

TDC

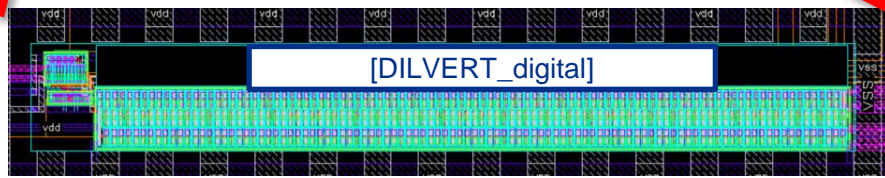


$\approx 250 \mu\text{m}$



$\approx 1000 \mu\text{m}$

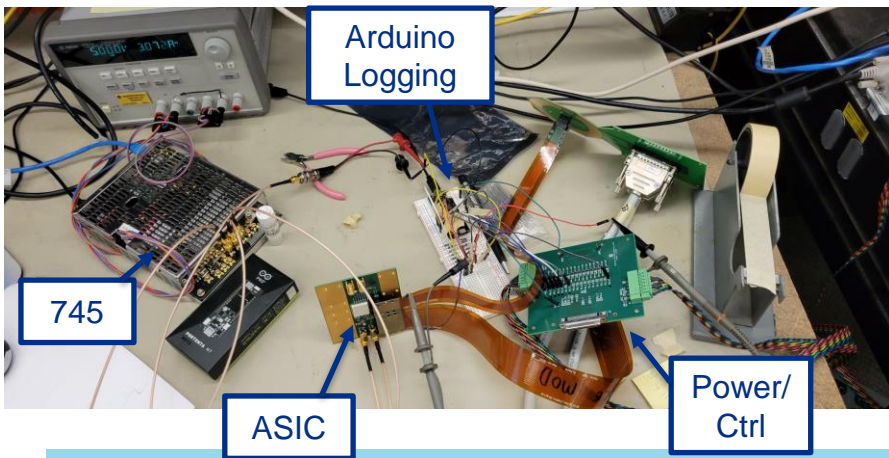
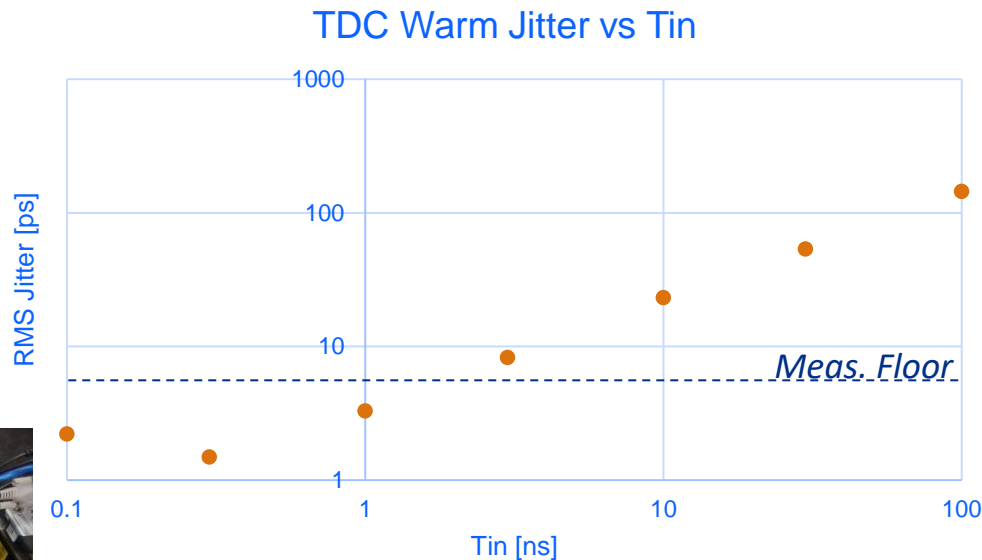
$\approx 20 \mu\text{m}$



$\approx 150 \mu\text{m}$

Preliminary Test Results (Room Temperature)

- Two pulses generated with Berkley Nucleonics 745-OEM (~ 5 ps rms jitter)
- RO runs slower and with more excess jitter than expected at room temperature.
- Cryo testing in progress.



Warm Simulation:

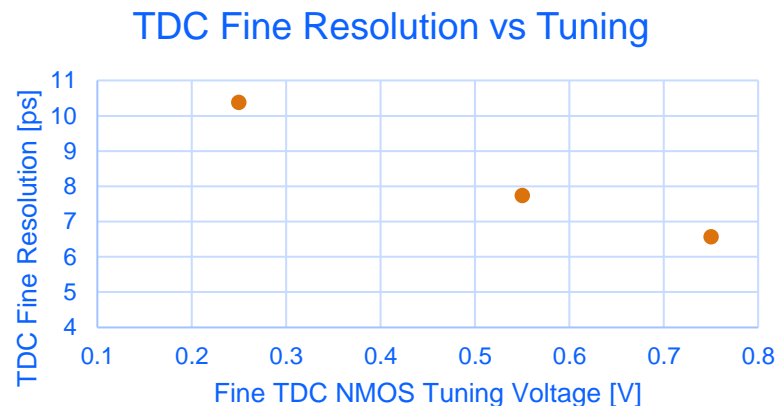
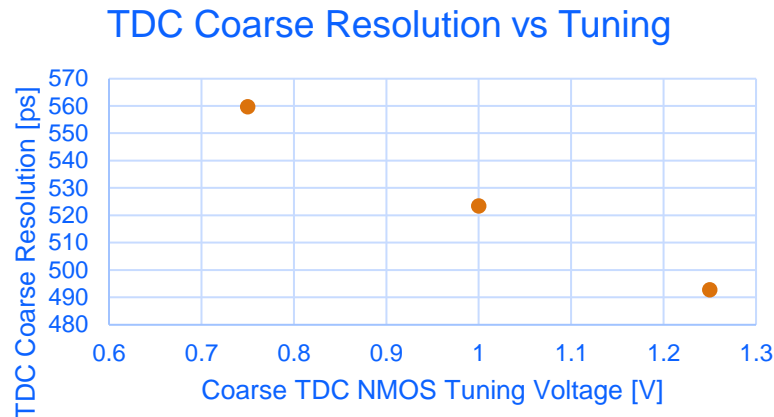
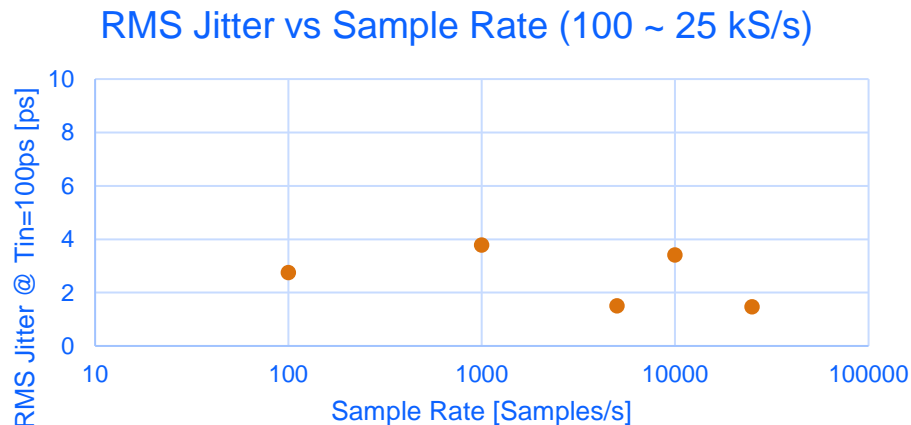
tR(coarse)	220.2 ps
tR(fine)	-6.0 ps
tOS	486.3 ps

Measurement:

tR(coarse)	524.4 ps
tR(fine)	-8.1 ps
tOS	-463.1 ps

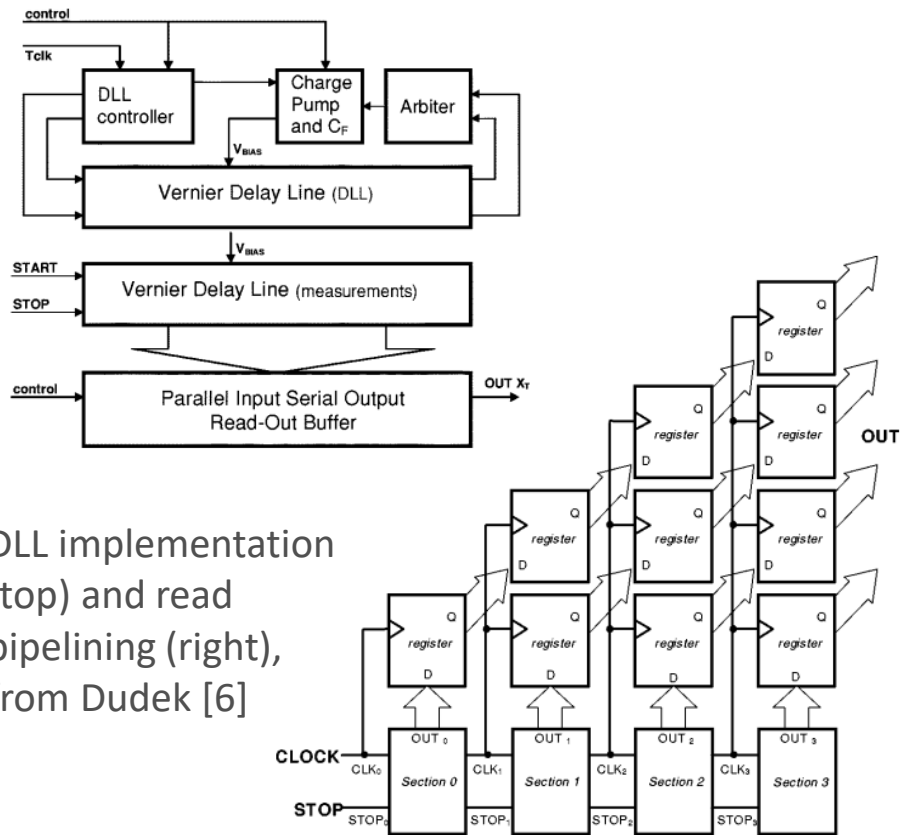
Performance vs Tuning Voltage & Sample Rate (Room Temperature)

- Demonstrated 25 kS/s, limited by readout chain.
- TDC Tuning slope: $\approx 134 \text{ ps/V}$ (coarse)
 $\approx 7.6 \text{ ps/V}$ (fine)



What's Next?

- Continue test chip characterization
- **Multi-Channel Prototype:** Power Saving in Ring Osc. / Clock Distribution
- **Full-Channel Integration** (SNSPDs, Comparators, etc.)
- Replica TDC for Delay-Locking
- High-Speed Read Pipelining
- Skewed Inverter Topologies



Thanks to Contributors

Fermilab ASIC Department:

D. Braga, F. Fahim, L. Dal Monte, P. Rubinov, P. Klabbers

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