

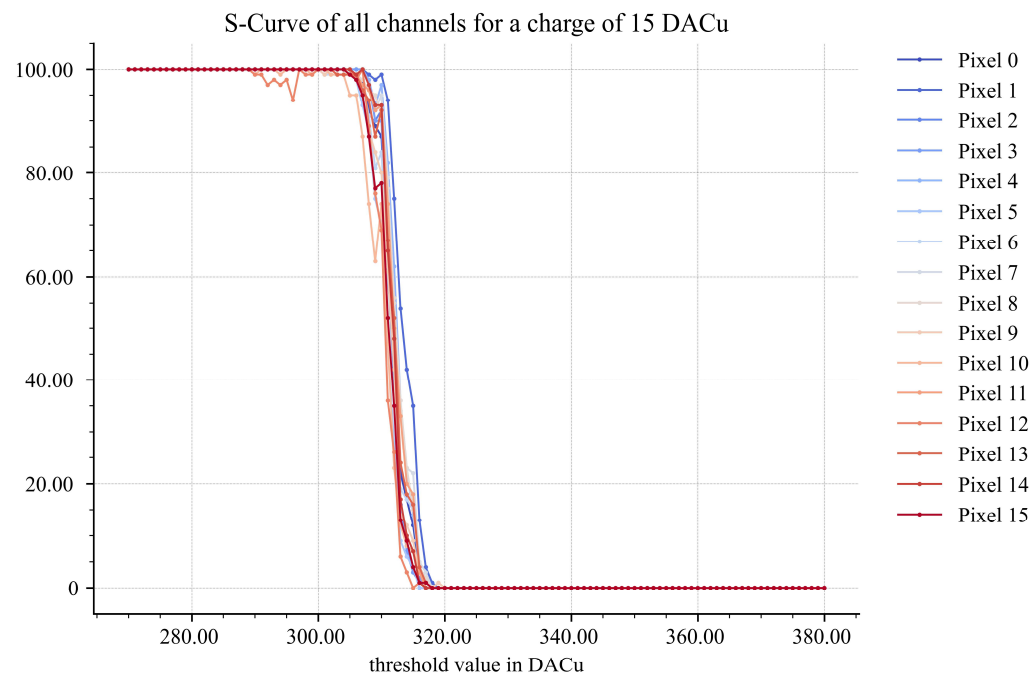
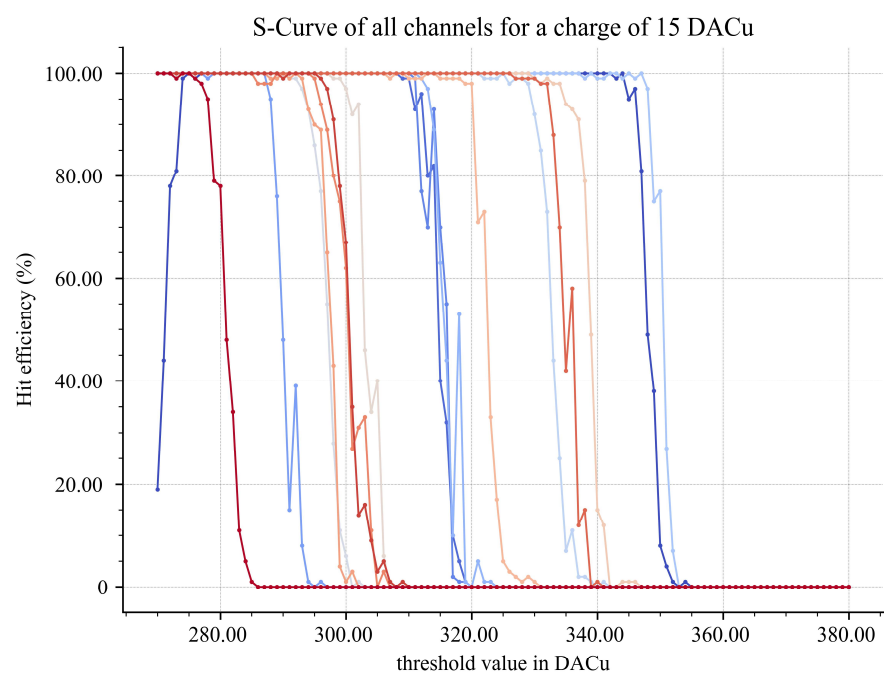
EICROC status and plans ePIC collaboration meeting BNL 21 jan 2026

F. Bouyjou, H. Chanal, F. Dulucq, M. El Berni, S. Extier, M. Firlej, T. Fiutowski, K. Guillosoy, F. Guilloux, M. Idzik, N. Kachkachi, B.-Y. Ki, C. de La Taille, J. Moron, D. Marchand, L. Royer, N. Seguin-Moreau, L. Serin, A. Shama, A. Soulier, K. Swientek, D. Thienpont, A. Verplancke

Organization for **M**icro-**E**lectronics desi**G**n and **A**pplications

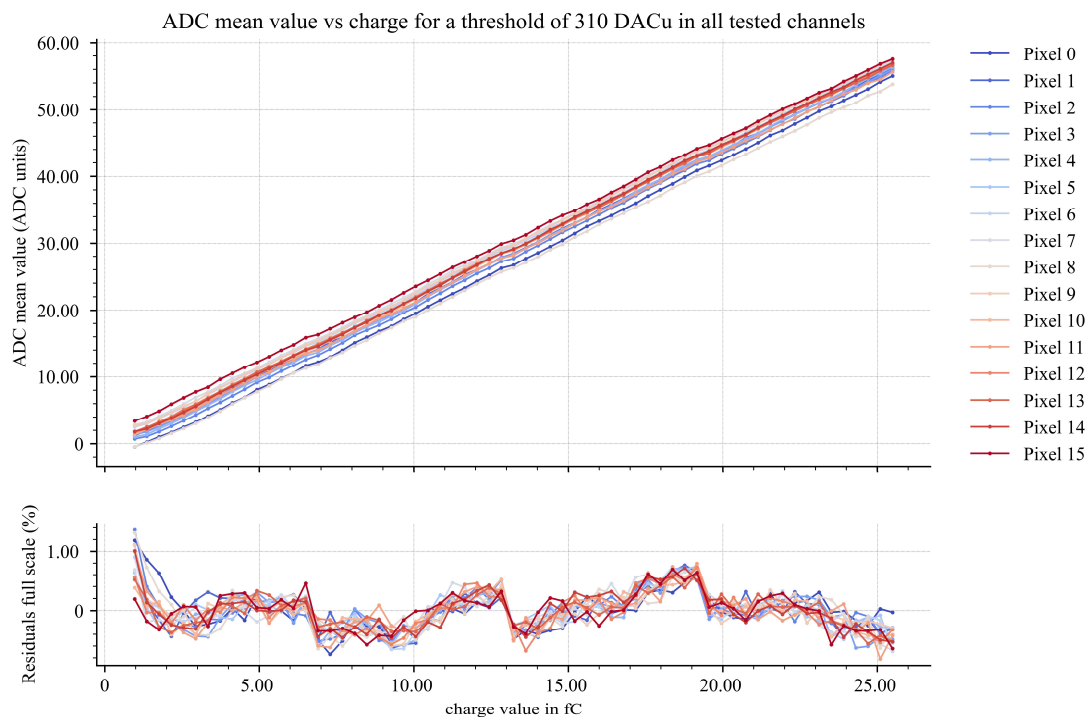
- 6-bit DAC for channel-wise threshold alignment

Correction of channels for a charge of 5 fC

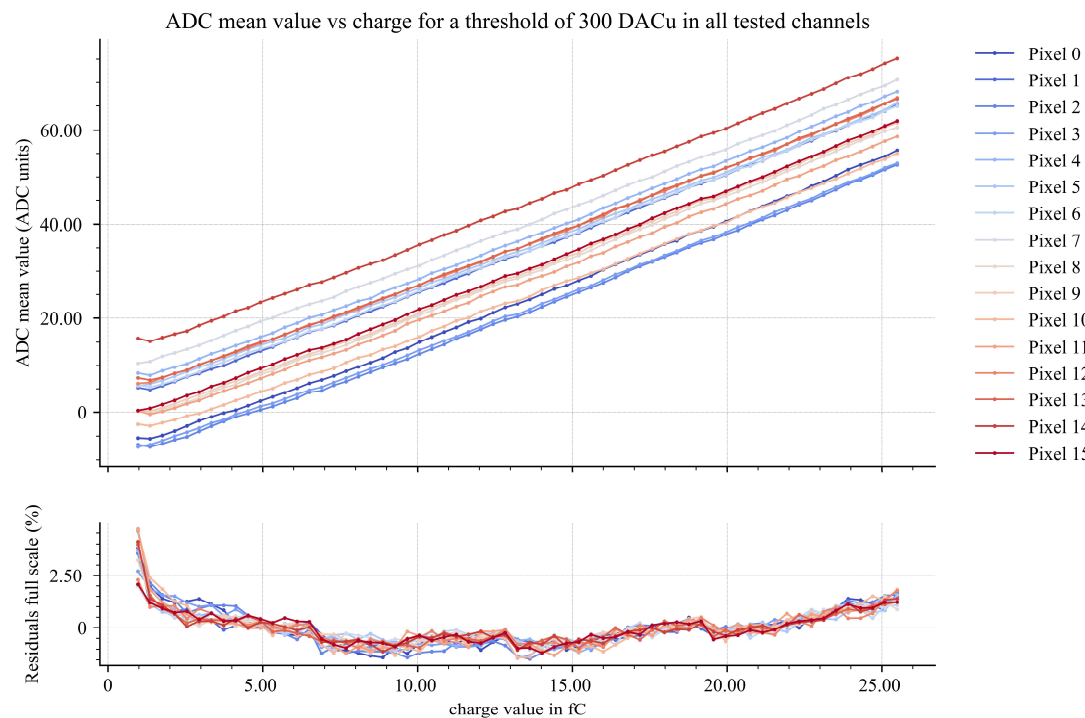


- 8 bit 40 MHz SAR ADC [Krakow]
 - Noise ~1 UADC but large common mode noise
 - Signal 50 UADC ~500 μm \Rightarrow 1 UADC ~10 μm

ADC mean & residuals – no sensor

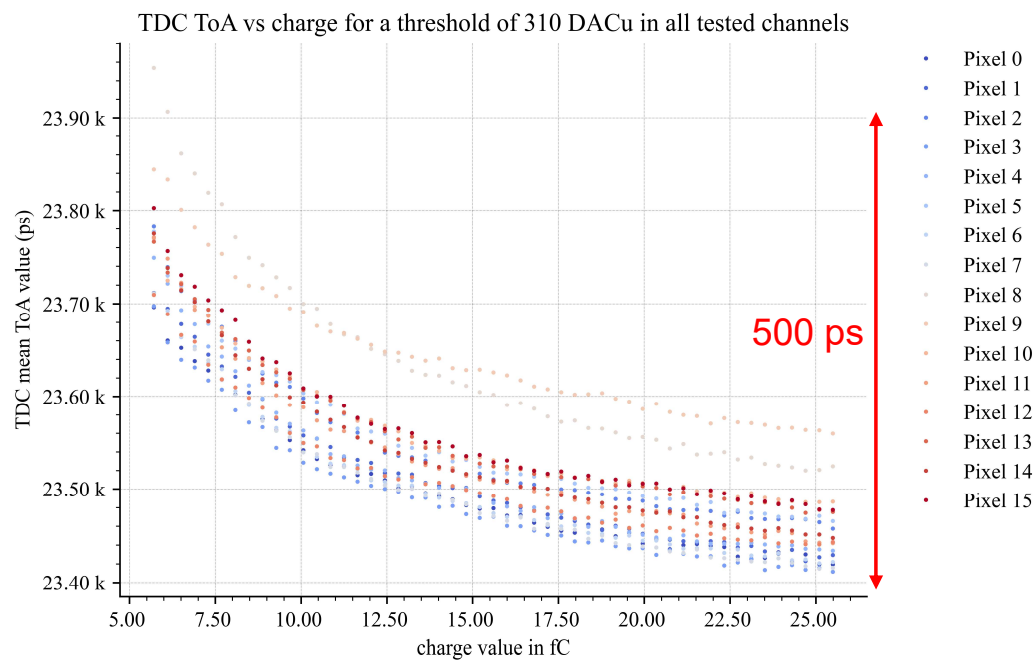


ADC mean & residuals – BNL Flipchip sensor

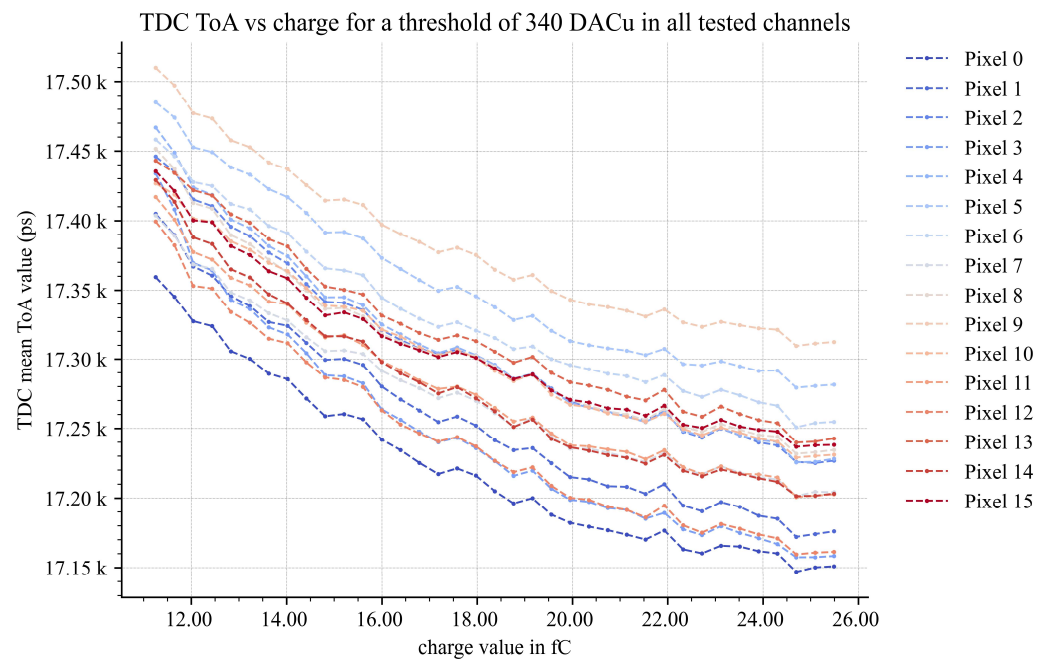


- Time walk measurement : ~400 ps

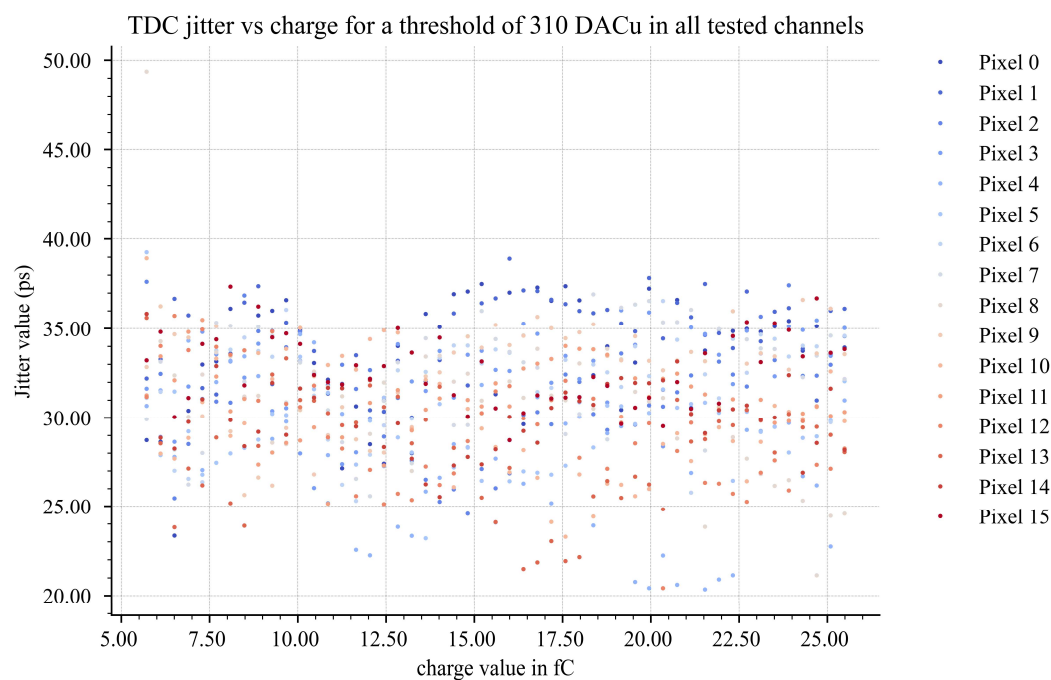
TDC ToA – no sensor



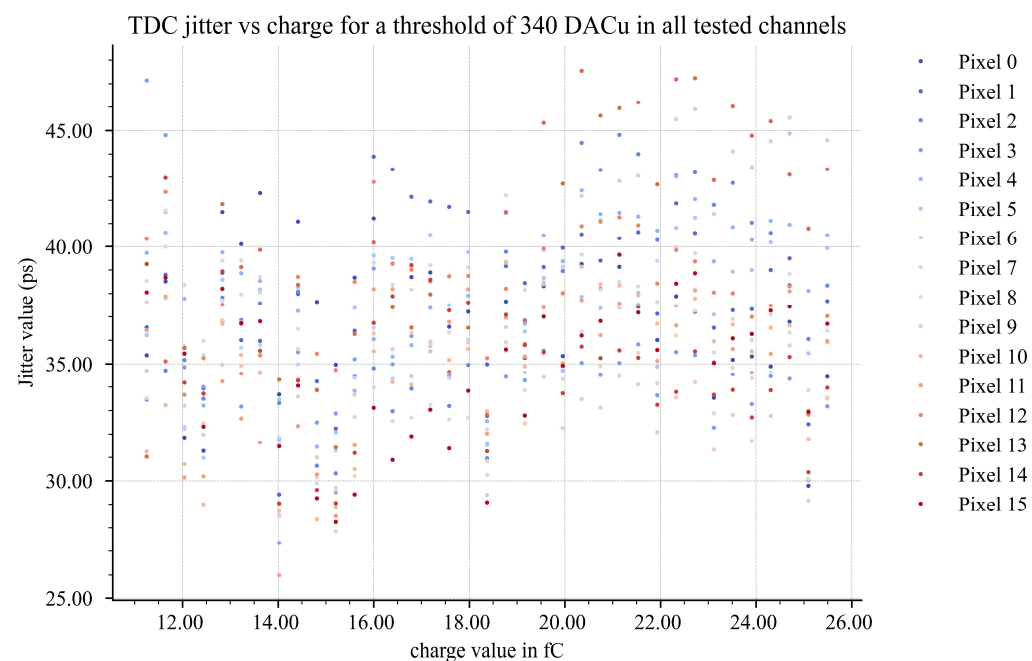
TDC ToA – BNL Flipchip sensor



TDC jitter – no sensor



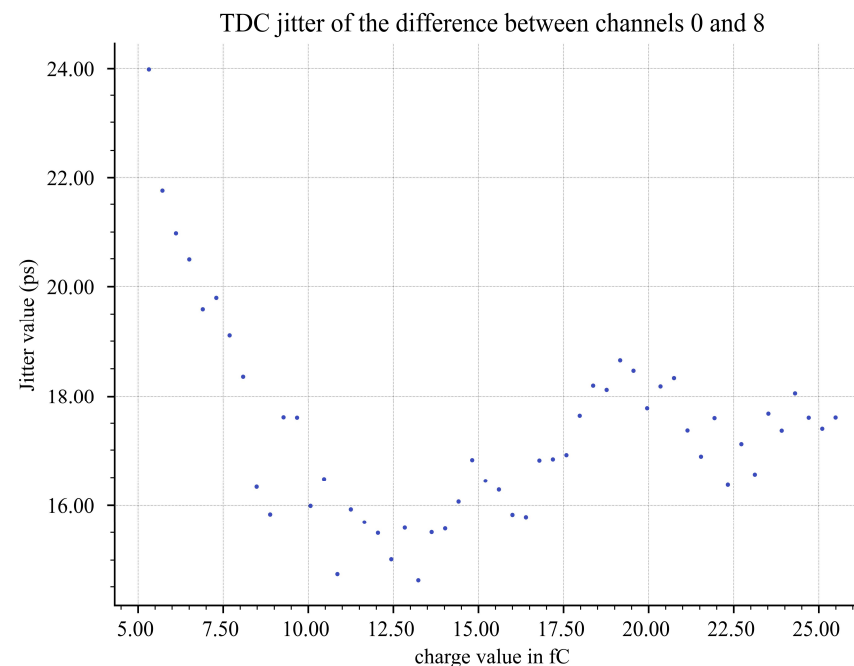
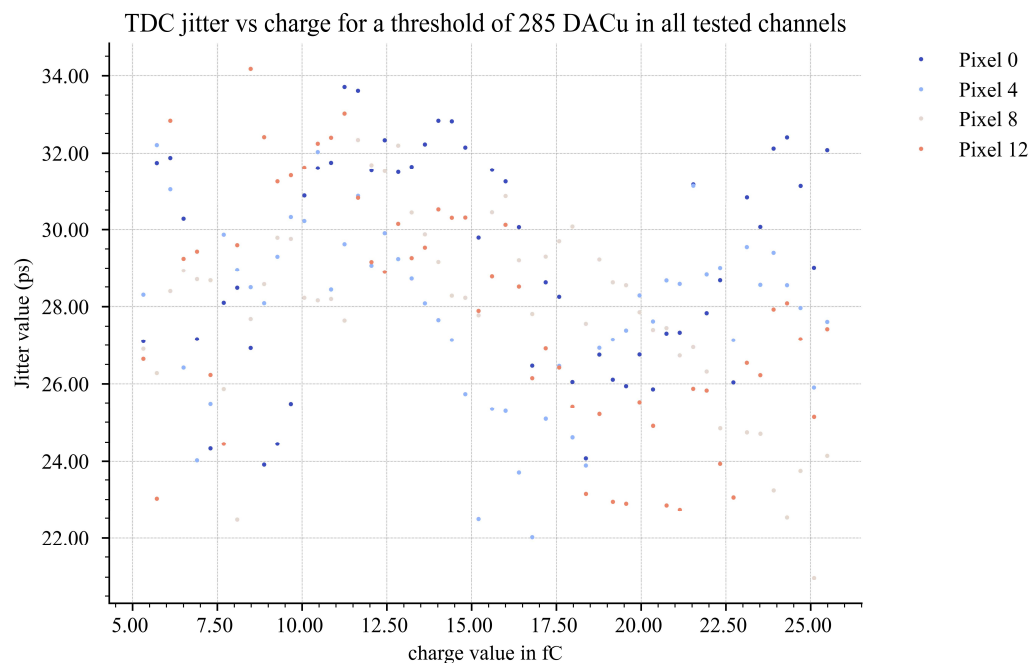
TDC jitter – BNL Flipchip sensor



Charge range degraded with sensor, still need to properly calibrate the TDC for each channel and remove correlated noise

EICROC0 measurements : TDC jitter

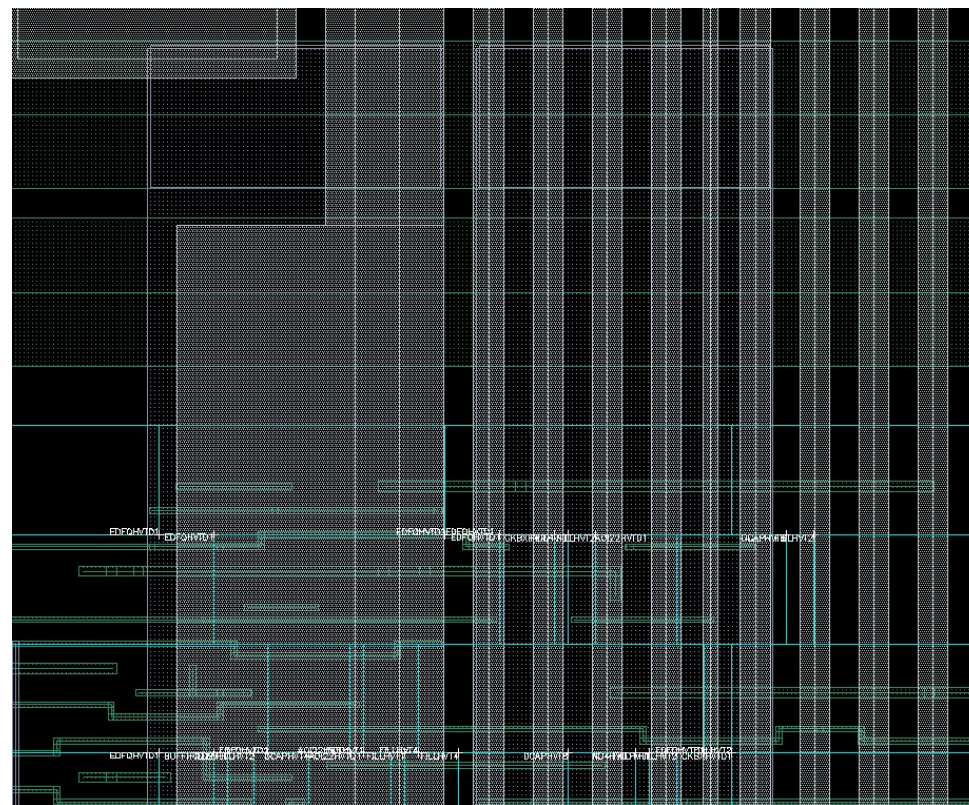
TDC jitter results for a simultaneous injection in pixels of line 0 (0, 4, 8, 12)



- We can also inject in channels simultaneously with the test pulse to then **remove the correlated noise** from the TDC measurements
- The plot represents the TDC jitter for the difference of two channels that we injected in simultaneously
- Performance approaching what is measured with the oscilloscope

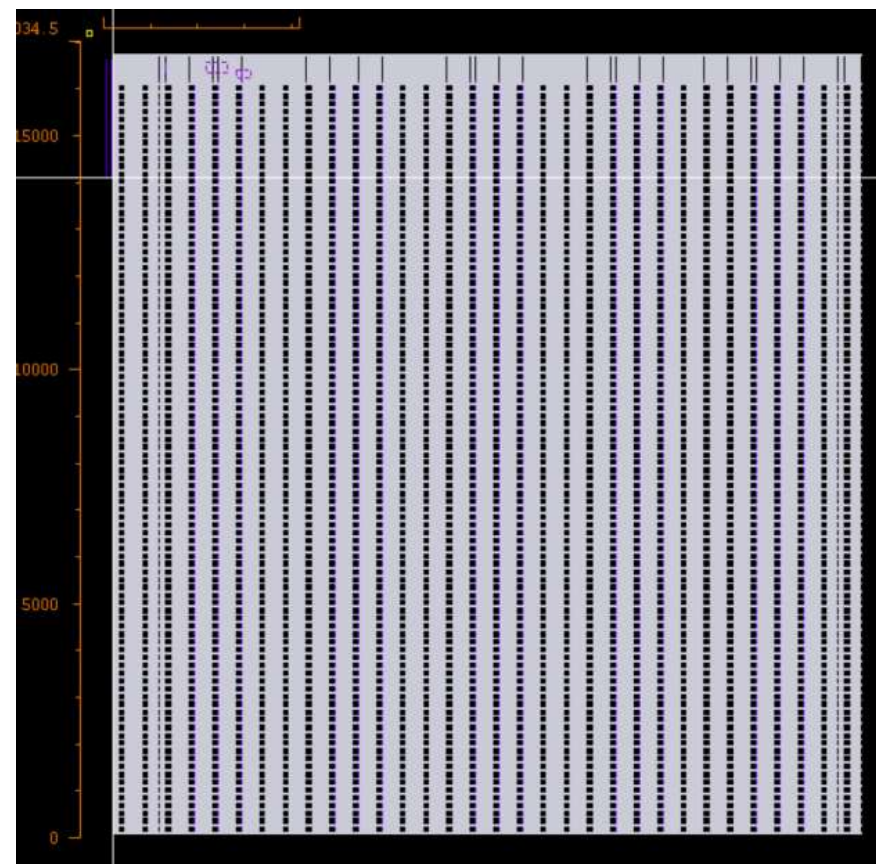
EICROC0 Coupling investigation

- Large coupling seen on the ADC data which was completely hiding the signal up to 20 fC
- Large coupling capacitance between local (pixel) digital block and preamp / shaper bias lines
- Layout improved by adding exclusion zones and a shield between bias lines and digital
- Done in EICROC0A



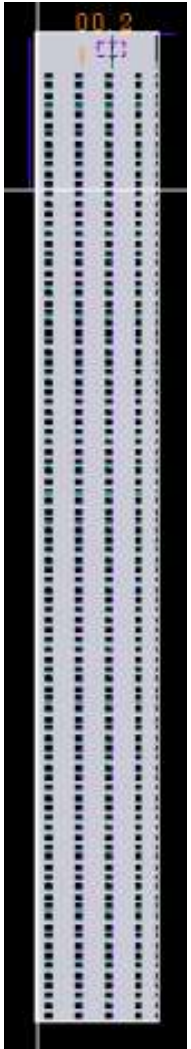
Bias / digital block layout in new EICROC ASICs

- 32x32 chip : final dimensions
- Goal :
 - test full-scale chip analog performance (IR drops)
 - Allow final sensor characterization
 - Test interface with DAQ : fast commands and 320 Mb/s data output
 - Progress on module/front-end boards
- Caveats :
 - Not (yet) zero-suppressed data
 - Still 2mW/ch
 - Still analog-on-top



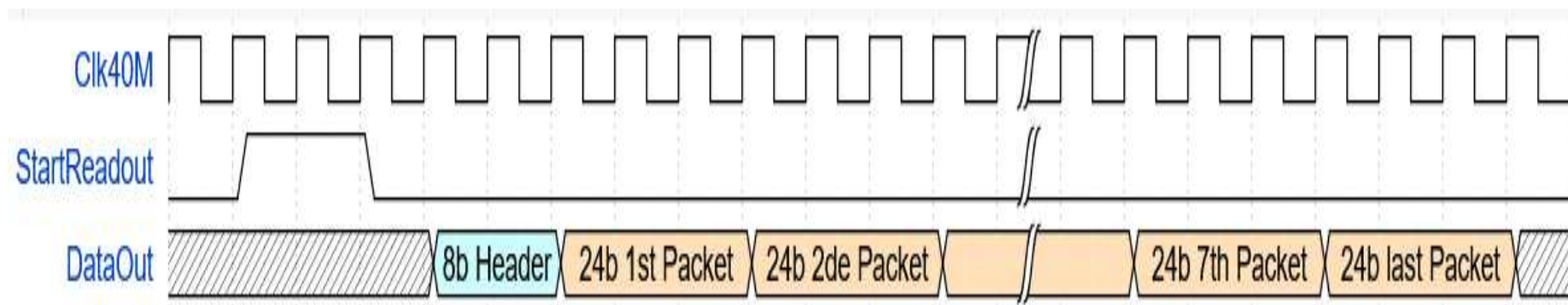
EICROC1 description

- Designed as 8 blocks of 4x32
 - Read out sequentially, like EICROC0
 - Pixel similar to EICROC0A
 - 8 CLPS outputs at 40 MHz
 - Clock tree for isochronous calib_pulse distribution
- Common balcony (End of column)
 - Biases and slow control
 - New : fast command decoder (clock, calib_pulse, enable_acquisition, start_readout)
 - Backup clock and cmd_pulse inputs available
 - New : 320 MHz serializer for 1 CLPS output



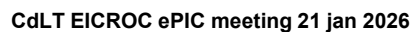
- Digital inputs
 - CLK_320 and FCMD (clock and fast commands) : CLPS
 - RSTb and RSTb_I2C : 1.2 V CMOS active low
 - SDA, SCL and 4b address : I2C slow control 1.2V CMOS
 - SEL_FCMD : 1.2V CMOS to select the backup inputs (below) (default fcmd)
 - Backup inputs : like EICROC0 (clk_160, cmd_pulse, en_acq, start_ro)
 - All unused inputs can be left floating (internal pull-ups/pull-downs)
- Digital outputs
 - 8 x 40 MHz data output (CLPS) by block of 4x32
 - one 320 MHz serialized data output (CLPS)
 - one trigger output (OR of all internal triggers) (CLPS)
 - One digital probe (1.2V CMOS) for debugging
- Analog outputs : for debugging (2 preamp outputs, analog probe, bias probe)

- Same as EICROC0 :
 - Data format : the **8 packets of 24 bits** represent 8 successive samples of the data. One packet of 24 formats is ordered as followed : packet[23]=0, packet[22:12]=TDC[10:0], packet[11:9]=0, packet[8:1]=ADC[7:0], packet[0]=hit



- Payload and readout speed
 - Each pixels provides 8 samples of 24 bits (12b TDC 12b ADC) at 40 MHz + header
 - $200 \text{ bits} * 25 \text{ ns} = 5 \text{ us/pixel}$
 - $128 \text{ pixels} * 5 \text{ us} = 640 \text{ us} \Rightarrow \sim 1.5 \text{ kHz maximum rate}$ per block of $4 \times 32 \sim 10 \text{ Hz/pixel}$
 - Same rate for full chip with serialized 320 MHz output
 - Spec for RPs is 15 Hz/pixel (max 28 Hz)
- Future improvements
 - Payload reduction to 12b TD C + 5 samples of 8b ADC = 64 bits $\Rightarrow \sim 30 \text{ Hz/pixel max rate}$
 - Zero-suppression : read only channels with hits and their (4-8) neighbours (for barycenters)
 - With 1 hit/column gives ~ 10 improvement $\Rightarrow \sim 300 \text{ Hz/pixel max rate}$
 - A derandomizer would also be needed

-
- The diagram illustrates the ATLAS detector layout. At the center is the interaction region, labeled 'ATLAS'. Surrounding it are the calorimeters, labeled 'CALOROCs'. The detector is divided into several sections, including the 'ATLAS' section (top), 'CALOROCs' section (bottom), and 'EICROCs' section (right). The layout shows the relative positions of the interaction region, calorimeters, and various subdetectors.



Future EICROC2

- What is missing in EICROC1 ?
 - Larger rate compatibility ?
 - **Derandomizer** for successive events (how many ?)
 - **Digital on Top** (DoT) design for digital power reduction and timing verification on large area : more powerful tools available (UVM)
 - Auto-trigger and zero_suppressed data
 - Power reduction if possible down to ~1 mW/ch
- Clermont Ferrand lab has joined to take responsibility for the DoT
 - Already participated in ATLAS ALTIROC
 - The design **and verification** should take ~1 year from now
- Technology choice ?
 - Depends mostly on rate estimates and complexity of digital

Architecture proposal

[A. Soulier et al.]

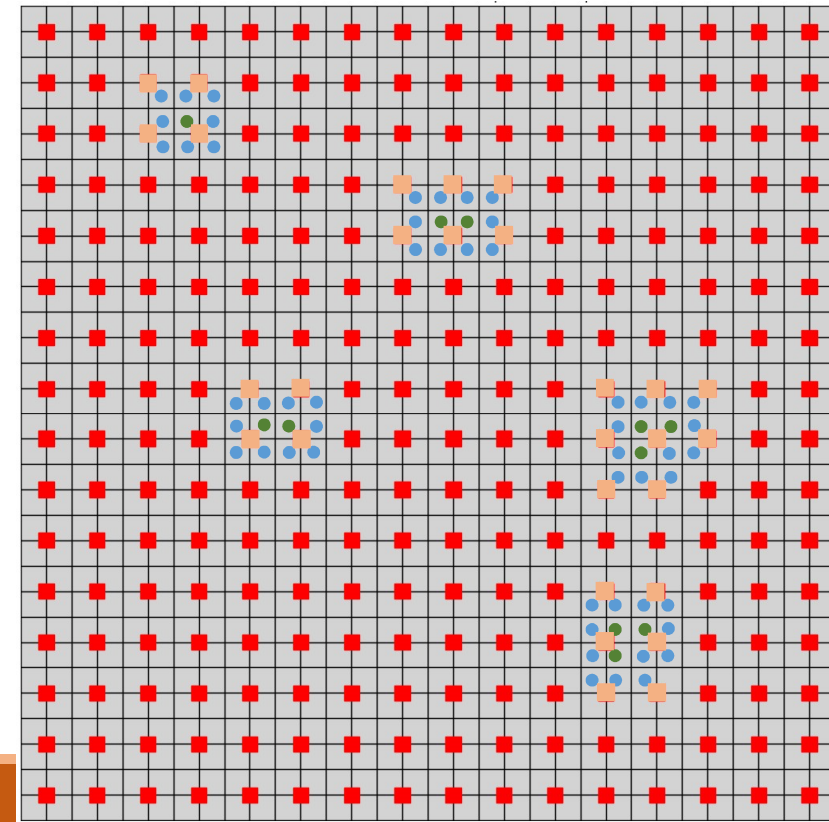


- Goals
 - Less digital logic possible inside pixel/matrix
 - Less signals possible between matrix and periphery
 - Triplication of sensitive circuits (configuration registers for example)
 - No triplication on data path
 - Complexity on periphery side, simple logic inside pixel
- Proposal : **Based on clusters of 4 pixels, controlled by the periphery**
 - Some logic shared at cluster level
 - Shared data bus between 2 columns

First iteration of this architecture is ready at RTL level, some debug in progress

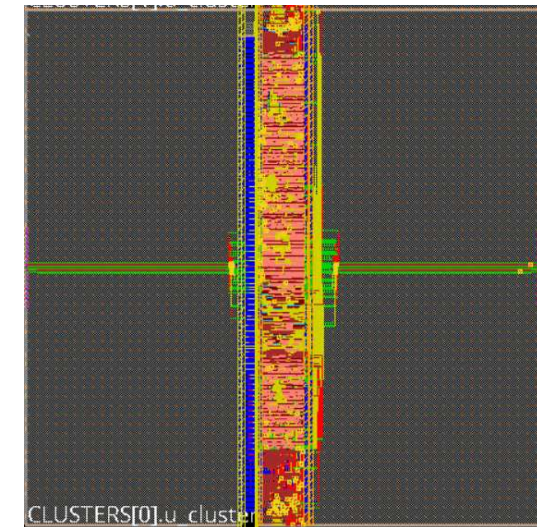
➔ Physical implementation can reveal new constraints, architecture may change

- Triggered Pixel (TP)
- Neighbor pixel (TL1)
- Busy cluster
- Cluster waiting for hit

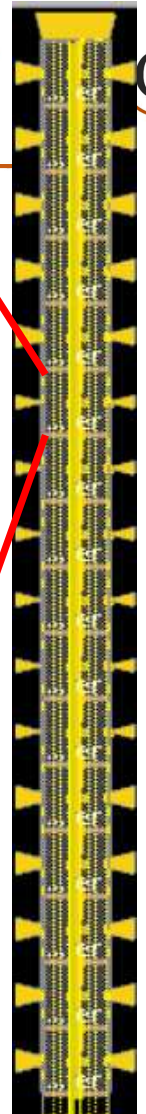


Physical implementation

- First physical implementation done in **130nm**
 - Check if we manage to fit digital readout inside pixel
- Preliminary layout of the cluster of 4 pixels and the column
- Triplication of cluster (probably need some refinement)
- Using a pixel_analog_dummy macro, not the real analog front-end
- **Need more time to validate that this architecture can fit using 130nm technology**



Cluster of 4 pixels



Double column

summary

- EICROC now reached full scale 32x32 with EICROC1
 - Important for analog performance and sensor test
 - Also allows to progress on system design
- Small scale EICROCs 4x4 still under test to improve performance and reduce power
 - Foresee also version in 65n in case digital does not fit in 130n
- EICROC2 now needs its digital design : ~1 year design time from now
 - With the most welcome help from Clermont Ferrand lab
 - Main driving specification is the hit rate : 50 Hz/pixel OK ?
- System tests are (as always) an important input for chip(s) finalization

- Layout Draft of E-ITO-TMVZ25-001

[illegible]

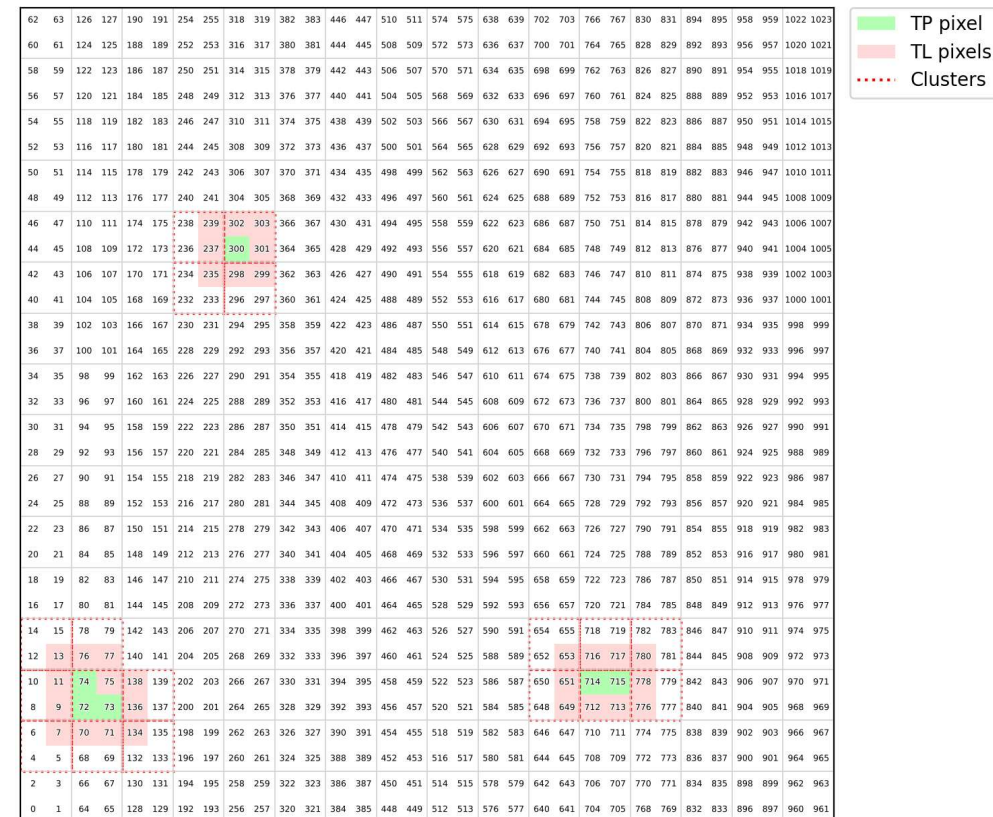
EICROC2 digital architecture

2 December 2025

Introduction

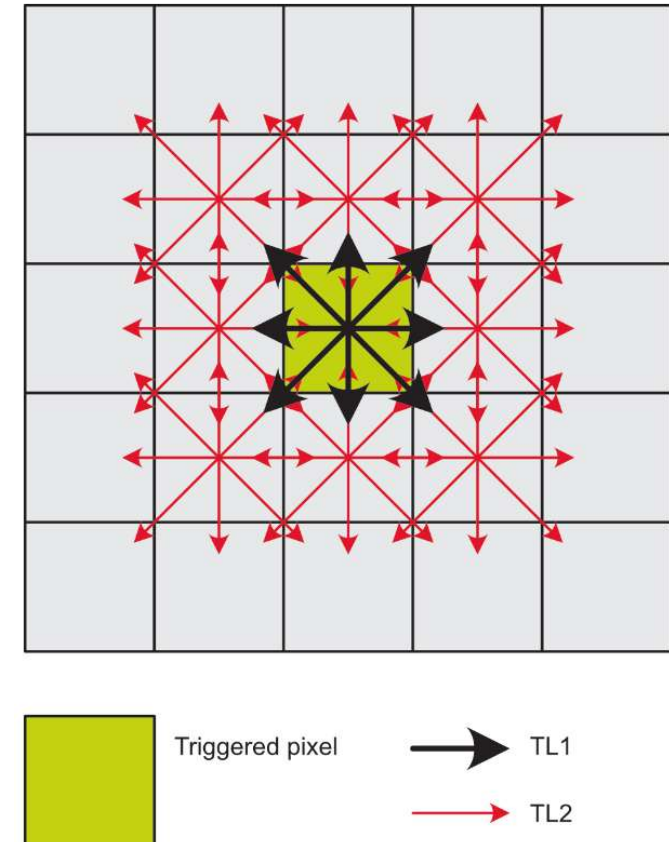


- AC LGAD sensors require to read pixel hits but also their neighbors due to charge sharing
- EICROC2 digital readout global specifications:
 - Must read neighboring pixels
 - Small area (can use 130nm techno if it is small enough)
 - Low power
 - Read fast enough to limit data loss due to dead time



Triggers for neighboring pixels

- **TP** : pixel receiving a hit = discriminator trigger, provide TDC and ADC data
- **TL1** : pixel receiving a trigger order from a TP, to read ADC data
- TL2 : pixel receiving a trigger order from a TL1, to read ADC data
- Implementation :
 - Only TL1 is implemented for now
 - ➔ Readout only the close neighbors
- Complexity of this solution will be when routing these signals at matrix level :
Each of the 1024 pixel has one output to its 8 neighbors, and 8 inputs from them



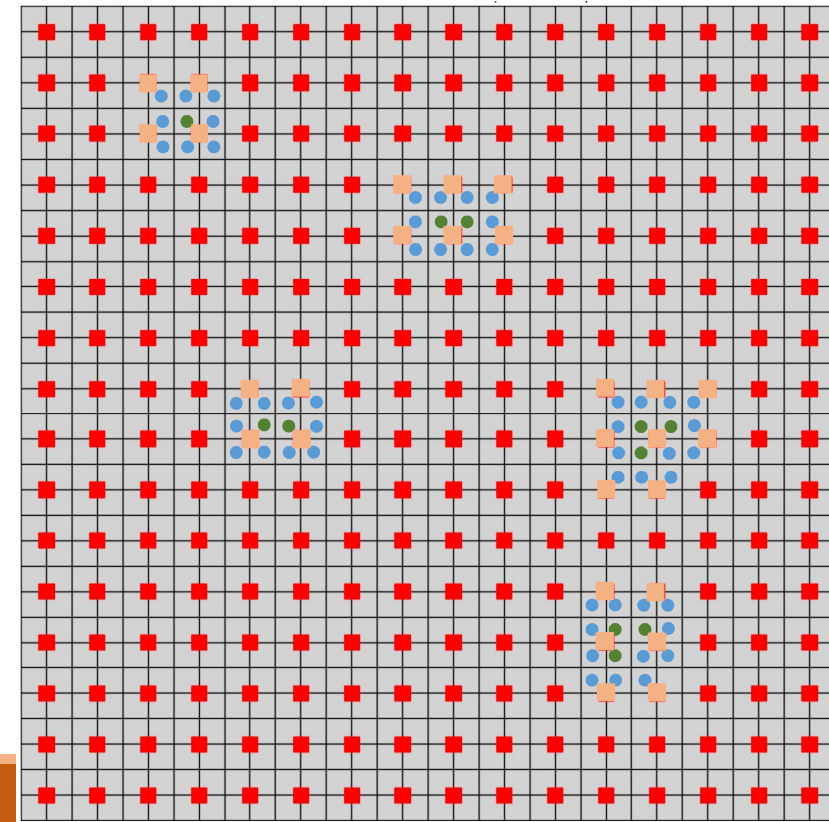
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What is the deadtime?



Dead time corresponds to the time a pixel is **blocked** and will not acquire new hits : pixel is **blind** during that time

- Analog dead time corresponds to the **ADC conversion time**
 - Digital dead time is the time to get the **data from the pixel to the periphery**
 - With the following hypothesis :
 - Event randomly distributed
 - Asynchronous system
- Efficiency can be defined as $R = \frac{1}{\lambda \delta + 1}$

Where λ is the event rate and δ the dead time

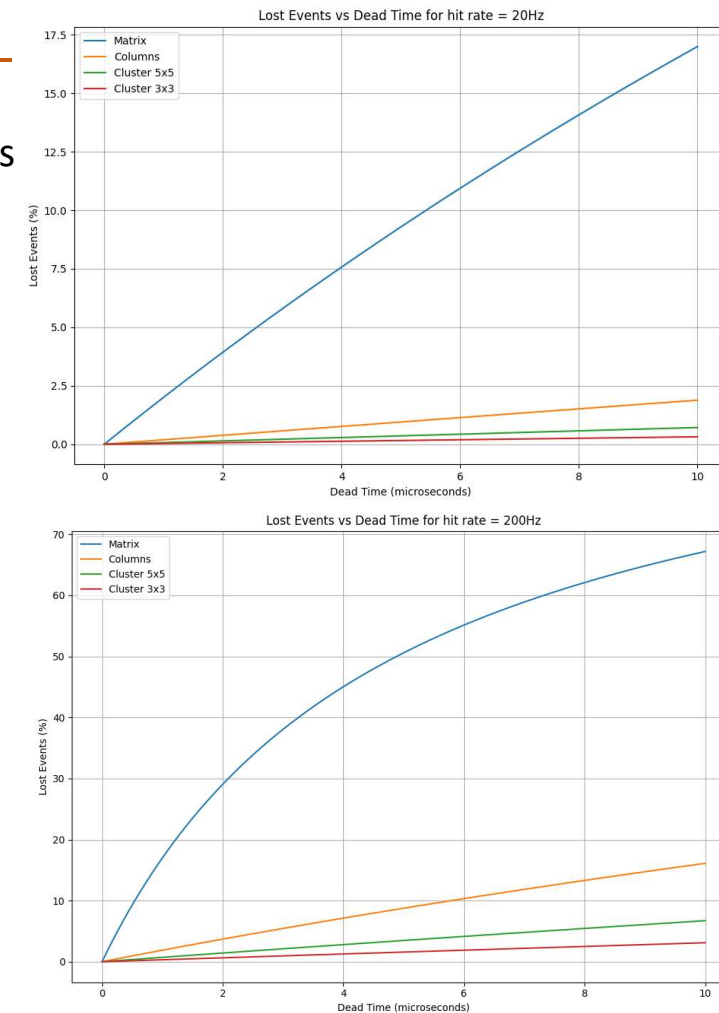
- From Alex Jentsch : « Event rate of 100Hz for the 10% pixels closest to the beam (beam halo), 10Hz for other pixels »
 - Event rate used for analyses is 20Hz per pixel as default, and 200Hz as worst case (considering double pixels hit)
- Dead time comes from simulation results of our architecture

Dead time

- Dead time also highly depends on the number of pixels blocked when a hit occurs
- Plots show how dead time impacts data loss for different blocking cases
- **Specifications :**
 ➔ Data loss due to dead time : **0.001% / 0.05%** (99.999 % / 99.95 % efficiency)
- **First estimation of data loss with such readout :**
 Considering an average analog dead time (ADC conversion time) of 400ns

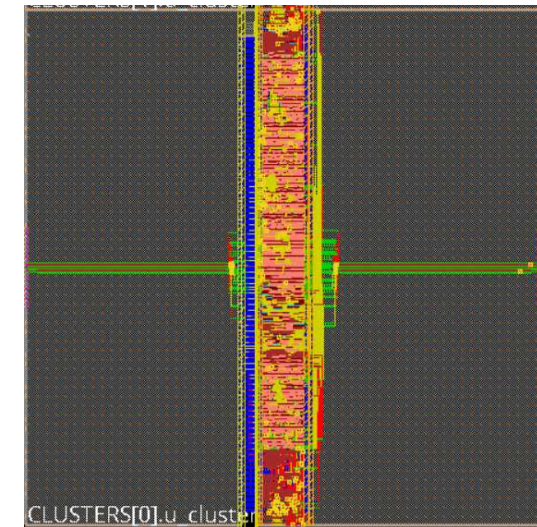
Data loss for Clusters, 3x3 readout	Rate = 20Hz	Rate = 200Hz
Default (1 hit = 1 TP, 8 TL1)	0.028%	0.28%
Multiple (1 hit = 2 TP, 10 TL1)	0.032%	0.32%
Worst case (1 hit = 4 TP, 12 TL1)	0.040%	0.40%

<0.05% data loss is achievable
0.001% data loss not possible for ADC with 800ns conversion time

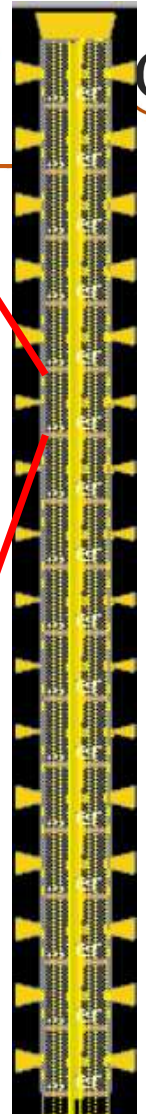


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Cluster of 4 pixels



Double column

Organisation and schedule

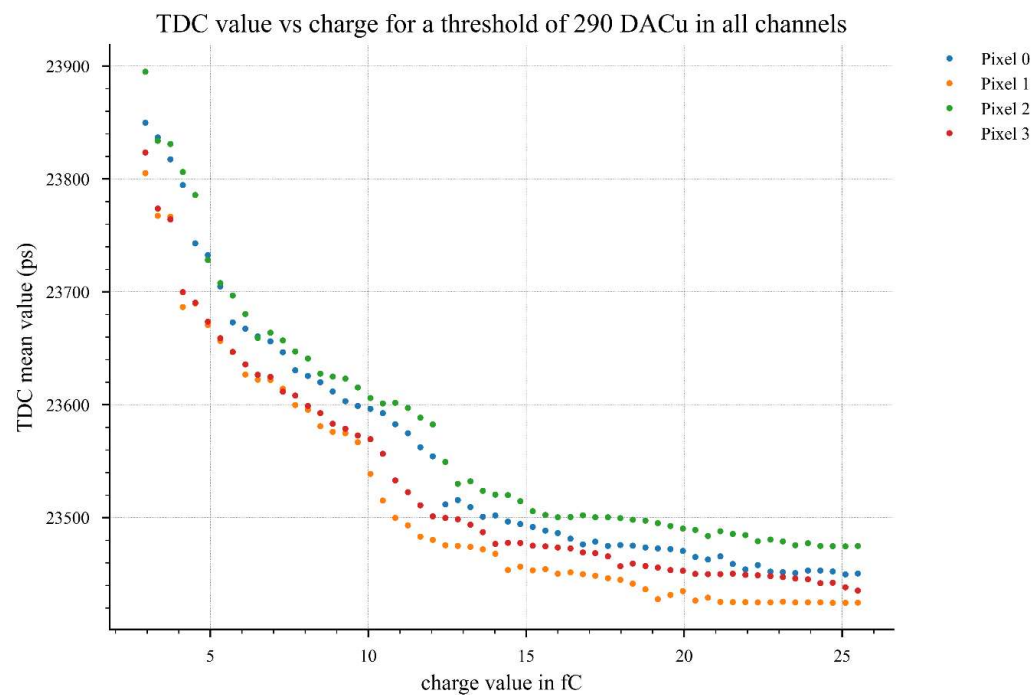


- LPCA team for EICROC :
 - 2 digital engineers (N. Kachkachi, A. Soulier)
 - 3 experts helping for specific points (H.Chanal for architecture, L.Royer for mixed-signal/system, N.Arveuf for UVM)
- Code and scripts shared on IN2P3 git
- Documentations shared on box IN2P3
 - EICROC spec : document gathering global specification related to the digital readout of the chip
 - Implementation spec : Detailed architecture explanation
- SOS server planned to be setup at Clermont : Centralized database for IP sharing
- Schedule planned:
 - First iteration of full digital RTL architecture including all blocks : Summer 2026
 - Technology choice : Early 2026
 - First iteration of full chip implementation : Autumn 2026

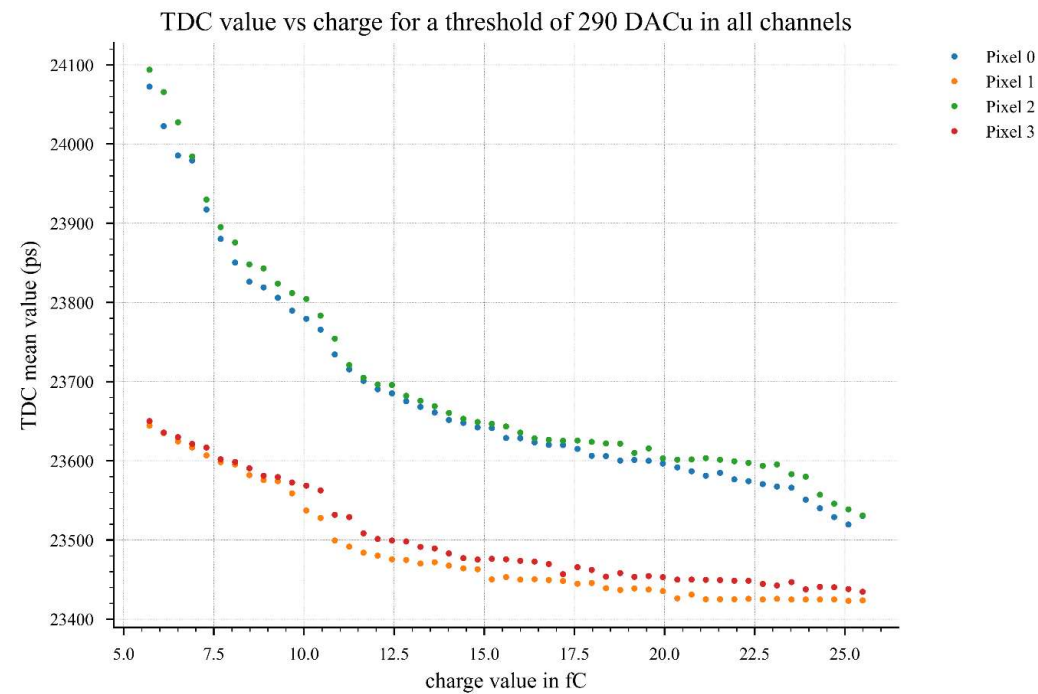
EICROC0 measurements : TDC ToA with no sensor

TDC ToA for $c_D = 00$ and $c_D = 11$

$c_D = 00$

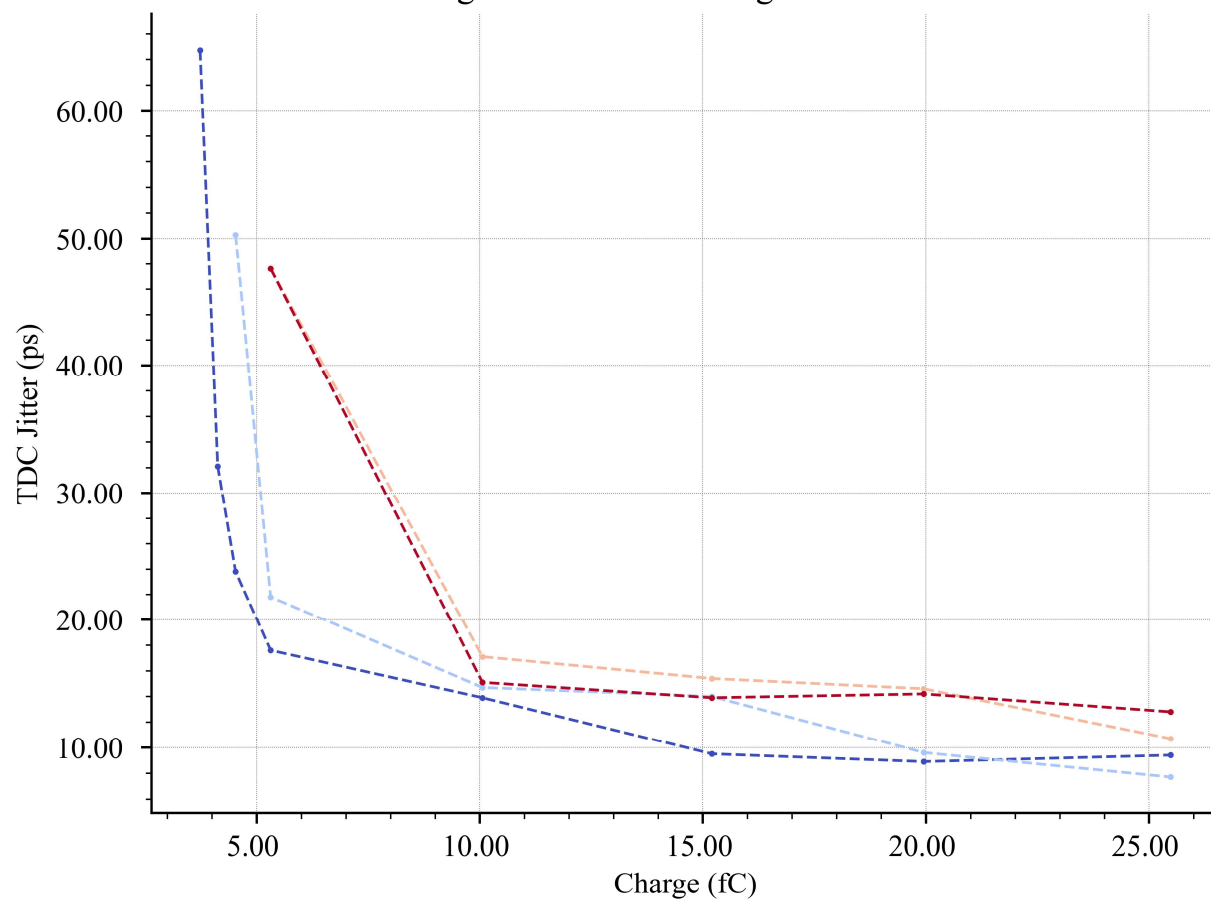


$c_D = 11$
Pixels 0 & 2



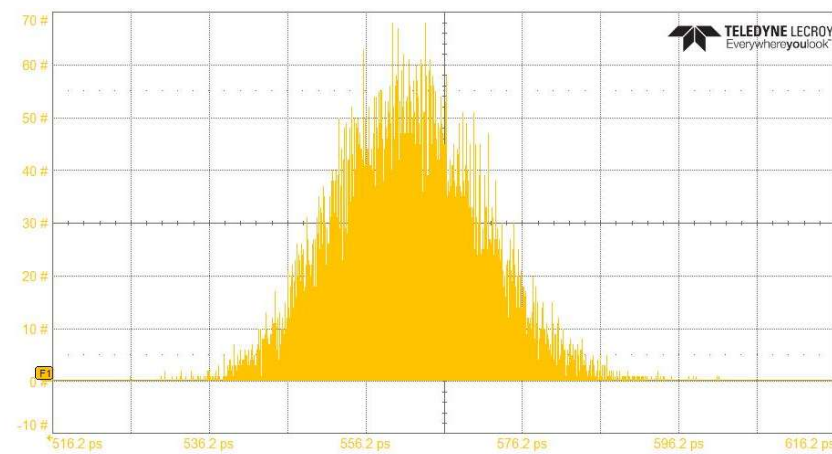
EICROC0 : discriminator jitter measured on probe

TDC Jitter vs charge for different configurations - EIC2.2 Channel 0



Measurements done with
the oscilloscope

- th280 & cd00
- th 290 & cd00
- th285 & cd11
- th295 & cd11



EICROC0 variants

- EICROC0A : same as EICROC0 with more testability
 - Each pixel can be by-passed by SC (clock is then turned off)
 - Buffers in SC to allow extension to 4x32
 - Larger dynamic range in pulse injection
- EICROC0B [Adrien]
 - Same preamp/discriminator/TDC/integrator
 - ADC and driver replaced by **peak sensing and Wilkinson ADC**
 - Currently ADC path ~ 1 mW/ch becomes 200 μ W/ch
- Fabricated with EICROC1 below

